

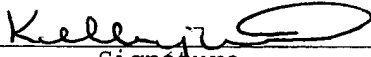
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APPLICATION FOR UNITED STATES LETTERS PATENT

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Title: EVALUATING AND OPTIMIZING ERROR-CORRECTING
CODES USING A RENORMALIZATION GROUP
TRANSFORMATION

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Evaluating and Optimizing Error-Correcting Codes using a Renormalization Group Transformation

Field of the Invention

The present invention relates generally to the field of error-correcting codes for
5 data storage and data transmission, and more particularly to evaluating and
optimizing error-correcting codes for intermediate length data blocks.

Background of the Invention

10 A fundamental problem in the field of data storage and communication is
generating optimal or near optimal error-correcting codes (ECC) for data of a
given block-length and transmission rate that can also be practically decoded. This
problem is now nearly solved for small block-lengths, e.g., blocks of length $N <$
100 bits, and for very large block-lengths, e.g., $N > 10^6$ bits. However, error-
15 correcting codes that are used in many applications, for example, wireless
communication, typically have block-lengths in an intermediate range, around $N =$
2000 bits. Generating optimal codes for these block-length remains a problem.

20 A large number of error-correcting codes are known for small block-lengths, many
of which are known to be optimal, or near optimal. As long as the block-length is
small enough, these ECC can be decoded practically and optimally using
maximum-likelihood decoders.

The problem of finding optimal codes for very large block-lengths has been essentially solved by parity-check codes defined by generalized parity check matrices. These types of codes were first described by R. G. Gallager, in "Low-density parity check codes," Vol.21, Research Monograph Series, MIT Press, 1963, but were not properly appreciated until recently. More recently, improved codes defined by sparse generalized parity check matrices have been described, such as turbocodes, irregular low-density parity check (LDPC) codes, Kanter-Saad codes, repeat-accumulate codes, and irregular repeat-accumulate codes.

These improved codes have three particularly noteworthy advantages. First, the codes can be decoded efficiently using belief propagation (BP) iterative decoding. Second, the performance of these codes can often be theoretically analyzed using a density evolution method, in an infinite-block-length limit. Third, by using a density evolution method, it can be demonstrated that these codes are nearly optimal codes. In the infinite-block-length limit, BP decoding of these codes decodes all data blocks that have a noise level below some threshold level, and that threshold level is often not far from the Shannon limit.

The preferred prior art way for generating improved codes has been to optimize codes for the infinite block-length limit using density evolution, and hope that a scaled-down version still results in a near optimal code. The problem with this method is that for $N < 10^4$, at least, the block-length is still noticeably far from the infinite-block-length limit. In particular, many decoding failures are found at noise levels far below the threshold level predicted by infinite block-length calculations. Furthermore, there may not necessarily even exist a way to scale down the codes derived from the density evolution method.

For example, the best known irregular LDPC codes, at a given rate in the $N \rightarrow \infty$ limit, often have variable nodes that should participate in hundreds or even thousands of parity checks, which makes no sense when the overall number of parity checks is 100 or less.

Density Evolution Method for a Binary Erasure Channel (BEC)

The density evolution method is simple for a binary erasure channel. A binary erasure channel is a binary input channel with three output symbols: 0, 1, and an erasure, which can be represented by a question mark “?” Because this method is important background for the method according to the invention, it is distinguished in greater detail.

Parity Check Codes

Linear block binary error-correcting codes can be defined in terms of a parity check matrix. In a parity check matrix A , the columns represent transmitted variable bits, and the rows define linear constraints or checks between the variable bits. More specifically, the matrix A defines a set of valid vectors or codewords z , such that each component of z is either 0 or 1, and

$$Az = 0, \tag{1}$$

where all multiplication and addition are modulo 2.

If the parity check matrix has N columns and $N-k$ rows, then the parity check defines an error correcting code of block-length N and transmission rate k/N , unless

some of the rows are linearly dependent, in which case some of the parity checks are redundant, and the code actually has a higher transmission rate.

As shown in Figure 1, there is a corresponding Tanner graph for each parity check matrix, see R. M. Tanner, "A recursive method to low complexity codes," IEEE Trans. Info. Theory, IT-27, pages 533-547, 1981. The Tanner graph 100 is a bipartite graph with two types of nodes: variable nodes i denoted by circles, and check nodes a denoted by squares. In the Tanner graph, each variable node is connected to the check node participating in the check for the variable node i .

For example, the parity check matrix

$$A = \begin{pmatrix} 1 & 1 & 0 & 1 & 0 & 0 \\ 1 & 0 & 1 & 0 & 1 & 0 \\ 0 & 1 & 1 & 0 & 0 & 1 \end{pmatrix} \quad (2)$$

is represented by the bipartite Tanner graph shown in Figure 1.

It should be understood, that in practical applications the graphs typically include thousands of nodes connected in any number of different ways, and containing many loops. Analyzing such graphs to determine optimal configurations is difficult.

Error-correcting codes defined by parity check matrices are linear. This means that each codeword is a linear combination of other codewords. In a check matrix, there are 2^k possible codewords, each of length N . For the example given the above, the

codewords are 000000, 001011, 010110, 011101, 100110, 101101, 110011, 111000. Because of the linearity property, any of the codewords are representative. For the purposes of analyzing a code, it is therefore normally assumed that the all-zeros codeword is transmitted.

5

Belief Propagation Decoding in the BEC

An input bit passes through the binary erasure channel as an erasure with probability x and is received correctly with probability $1 - x$. It is important to note that the BEC never flips bits from 0 to 1, or vice versa. If all-zeros codewords are transmitted, the received word must consist entirely of zeros and erasures.

10

The receiver uses a belief propagation (BP) decoder to decode the input bits by passing discrete messages between the nodes of the Tanner graph. A message m_{ia} is sent from each variable node i to each check node a connected to it. The message represents the state of the variable node i . In general, the message can be in one of three states: 1, 0, or ?, but because the all-zeros codeword is always transmitted, the possibility that m_{ia} has value 1 can be ignored.

15

Similarly, there is a message m_{ai} sent from each check node a to all the variable nodes i connected to the check node. These messages are interpreted as directives from the check node a to the variable node i about what state the variable node should be in. This message is based on the states of the other variable nodes connected to the check node. The check-to-bit messages can, in principle, take on the values 0, 1, or ?, but again only the two messages 0 and ? are relevant when the all-zeros codeword is transmitted.

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In the BP decoding process for the BEC, a message m_{ia} from a variable node i to a check node a is equal to a non-erasure received message because such messages are always correct in the BEC, or to an erasure when all incoming messages are erasures. A message m_{ai} from a check node a to a variable node i is an erasure when any incoming message from another node participating in the check is an erasure; otherwise it takes on the value of the binary sum of all incoming messages from other nodes participating in the check.

BP decoding is iterative. The messages are initialized so that all variable nodes that are not erased by the channel send out messages equal to the corresponding received bit, and all other messages are initially erasures. Iterating the BP message process converges eventually to stationary messages because convergence of BP decoding is guaranteed for the particularly simple BEC, though not for other channels. The final decoded value of any erased variable node is just the value of any non-erasure message coming into that node, unless there is no incoming non-erasure message. In this case, the BP decoding process terminates and fails to decode the particular variable node.

Density Evolution

The average probability of failure for BP decoding over many blocks is now considered. A real number p_{ia} , which represents the probability that the message m_{ia} is an erasure, is associated with each message m_{ia} . Similarly, a real number q_{ai} , which represents the probability that the message m_{ai} is an erasure, is associated with each message m_{ai} . In the density evolution method, probabilities p_{ia} and q_{ai} are

determined in a way that is exact, as long as the Tanner graph representing the error-correcting code has no loops.

The equation for p_{ia} is

$$p_{ia} = x \prod_{b \in N(i) \setminus a} q_{bi}, \quad (3)$$

where $b \in N(i) \setminus a$ represents all check nodes directly connected to a neighboring variable node i , except for check node a . This equation can be derived from the fact that for a message m_{ia} to be an erasure, the variable node i must be erased in transmission, and all incoming messages from other checks are erasures as well. Of course, if the incoming messages are correlated, then this equation is not correct. However, in a Tanner graph with no loops, each incoming message is independent of all other messages.

Similarly, the equation

$$q_{ai} = 1 - \prod_{j \in N(a) \setminus i} (1 - p_{ja}) \quad (4)$$

can be derived from the fact that a message q_{ai} can only be in a 0 or 1 state when all incoming messages are in either a zero or one state.

The density evolution equations (3) and (4) can be solved by iteration. A good initialization is $p_{ia} = x$ for all messages from variable nodes to check nodes and $q_{ai} = 0$ for all messages from check nodes to variable nodes, as long as the iteration begins with the q_{ai} messages. The BEC density evolution equations ultimately converge. This can be guaranteed for codes defined in graphs without loops. It is possible to determine b_i , which is the probability of a failure to decode at variable node i , from the formula

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$$b_i = x \prod_{a \in N(i)} q_{ai} \quad (5)$$

Exact Solution of a Small Code

- 5 As stated above, the density evolution equations (3, 4, and 5) are exact when the code has a Tanner graph representation without loops.

Consider the error-correcting code defined by a parity check matrix

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$$A = \begin{pmatrix} 1 & 1 & 0 & 0 \\ 0 & 1 & 1 & 1 \end{pmatrix} \quad (6)$$

and represented by a corresponding Tanner graph shown in Figure 2. This code has four codewords: 0000, 0011, 1101, and 1110. If the 0000 message is transmitted, then there are sixteen possible received messages: 0000, 000?, 00?0, 00??, 0?00, and so on. The probability of receiving a message with n_e erasures is $x^{n_e}(1-x)^{4-n_e}$.

- 15 Messages might be partially or completely decoded by the BP decoder; for example the received message ?00? is fully decoded to 0000, but the message 0??? is only partially decoded to 00??, because there is not enough information to determine whether the transmitted codeword was actually 0000 or 0011.

- 20 It is easy to determine the exact probability that a given bit remains an erasure after decoding by summing over the sixteen possible received messages weighted by their probabilities. For example, the first bit is only decoded as an erasure when one of the following messages is received: ???0, ??0?, or ????, so the correct probability that the first bit is not decoded is $2x^3(1-x) + x^4 = 2x^3 - x^4$.

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If the focus is on the last bit, then the message is decoded, unless one of the following messages is sent: 00??, 0???, ?0??, ??0? or ????. Therefore, the overall probability that the fourth bit is not decoded is $x^2(1-x)^2 + 3x^3(1-x) + x^4 = x^2 + x^3 - x^4$. In the density evolution method, the values for the following variables:

$$p_{11}, p_{21}, p_{22}, p_{32}, p_{42}, q_{11}, q_{12}, q_{22}, q_{23}, q_{24}, b_1, b_2, b_3, b_4$$

are determined by equations

$$p_{11} = x \quad (7)$$

$$p_{21} = xq_{22} \quad (8)$$

$$p_{22} = xq_{12} \quad (9)$$

$$p_{32} = x \quad (10)$$

$$p_{42} = xq_{11} = p_{21} \quad (11)$$

$$q_{12} = p_{11} \quad (12)$$

$$q_{22} = 1 - (1 - p_{32})(1 - p_{42}) \quad (13)$$

$$q_{23} = 1 - (1 - p_{22})(1 - p_{42}) \quad (14)$$

$$q_{24} = 1 - (1 - p_{22})(1 - p_{32}) \quad (15)$$

and

$$b_1 = xq_{11} \quad (16)$$

$$b_2 = xq_{12}q_{22} \quad (17)$$

$$b_3 = xq_{23} \quad (18)$$

$$b_4 = xq_{24} \quad (19)$$

Solving these equations yields

$$p_{11} = x \quad (20)$$

$$p_{21} = 2x^2 - x^3 \quad (21)$$

$$p_{22} = x^2 \quad (22)$$

$$p_{32} = x \quad (23)$$

$$p_{42} = x \quad (24)$$

$$5 \quad p_{11} = 2x^2 - x^3 \quad (25)$$

and

$$q_{12} = x \quad (26)$$

$$q_{22} = 2x - x^2 \quad (27)$$

$$q_{23} = x + x^2 - x^3 \quad (28)$$

$$10 \quad q_{24} = x + x^2 - x^3 \quad (29)$$

and

$$b_1 = 2x^3 - x^4 \quad (30)$$

$$b_2 = 2x^3 - x^4 \quad (31)$$

$$b_3 = x^2 + x^3 - x^4 \quad (32)$$

$$15 \quad b_4 = x^2 + x^3 - x^4 \quad (33)$$

Examining the results for b_1 and b_4 indicates that the density evolution solution agrees exactly with the direct approach for this code.

20 The Large Block-Length Limit

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If all local neighborhoods in the Tanner graph are identical, the density evolution equations can be simplified. For example, if each variable node i is connected to d_v parity checks, and each check node a is connected to d_c variable nodes, then all the

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 p_{ia} are equal to the same value p , all the q_{ai} are equal to the same value q , and all the b_i are equal to the same value b . Then,

$$p = xq^{d_v-1} \quad (34)$$

$$q = 1 - (1 - p)^{d_c-1} \quad (35)$$

5 and

$$b = xq^{d_v} \quad (36)$$

which are the density evolution equations for (d_v, d_c) regular Gallager codes, valid in the $N \rightarrow \infty$ limit. A regular Gallager code is a sparse random parity check matrix characterized by the restriction that each row has exactly d_c ones in it, and each column contains exactly d_v ones.

The intuitive reason that these equations are valid, in the infinite block-length limit, is that as $N \rightarrow \infty$, the size of typical loops in the Tanner graph of a regular Gallager code go to infinity, so all incoming messages to a node are independent, and a regular Gallager code behaves as a code defined on a graph without loops. Solving equations (34 and 35) for specific values of d_v and d_c yields a solution that is $p = q = b = 0$, below a critical erasure limit of x_c . This means that decoding is perfect. Above x_c , b has a non-zero solution, which correspond to decoding failures. The value x_c is easy to determine numerically. For example, if $d_v = 3$ and $d_c = 5$, then $x_c \approx 0.51757$.

These density evolution calculations can be generalized to irregular Gallager codes, or other codes like irregular repeat-accumulate codes which have a finite number of different classes of nodes with different neighborhoods. In this generalization, one can derive a system of equations, typically with one equation for the messages leaving each class of node. By solving the system of equations,

one can again find a critical threshold x_c , below which decoding is perfect. Such codes can thus be optimized, in the $N \rightarrow \infty$ limit, by finding the code that has maximal noise threshold x_c .

- 5 Unfortunately, the density evolution method is erroneous for codes with finite block-lengths. One might think that it is possible to solve equations (3 and 4) for any finite code, and hope that ignoring the presence of loops is not too important a mistake. However, this does not work out, as can be seen by considering regular Gallager codes. Equations (3, 4, and 5) for a finite block-length regular Gallager
- 10 code have exactly the same solutions as one would find in the infinite-block-length limit, so one would not predict any finite-size effects. However, it is known that the real performance of finite-block-length regular Gallager codes is considerably worse than that predicted by such a naive method.
- 15 Therefore, there is a need for a method to correctly evaluate finite length error-correcting codes that do not suffer from the problems of the prior art methods.

Summary of the Invention

- 20 The present invention provides a method for evaluating an error correcting code for a data block of a finite size in a binary erasure channel or an additive white Gaussian noise channel. An error-correcting code is defined by a generalized parity check matrix, wherein columns represent variable bits and rows represent parity bits. In a notation of the matrix, “hidden” variable bits which are not transmitted
- 25 through the channel are represented by a bar over the corresponding column of the generalized parity check matrix. The generalized parity check matrix is represented

as a bipartite graph. A single node in the bipartite graph is iteratively renormalized until the number of nodes in the bipartite graph is less than a predetermined threshold.

- 5 During the iterative renormalization, a particular variable node is selected as a target node, and a distance between the target node and every other node in the bipartite graph is measured. Then, if there is at least one “leaf” variable node, renormalize a leaf variable node farthest from the target node, otherwise, renormalize a leaf check node farthest from the target node, and otherwise
10 renormalize a variable node farthest from the target node and having fewest directly connected check nodes. Leaf nodes are connected to only one other node in the graph.

- When the number of nodes in the graph is less than the predetermined threshold,
15 the decoding failure rate for the target node is determined exactly.

- The present invention provides a method for optimizing error-correcting codes by searching for the error-correcting code of a specified data block size and transmission rate with the best performance in terms of decoding failure as a
20 function of noise. The decoding failure rates for transmitted variable bits are used to guide the search for an optimal code.

Brief Description of the Drawings

Figure 1 is a prior art bipartite graph representing an error-correcting code including a loop;

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Figure 2 is a prior art bipartite graph representing a simple prior art error-correcting code;

Figures 3a-e are bipartite graphs renormalized according to the invention;

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Figure 4 is a bipartite graph to be renormalized according to the invention;

Figure 5 is a bipartite graph with loops to be renormalized;

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Figure 6 shows an expansion of a bipartite graph to be renormalized;

Figure 7 is a bipartite graph with a loop representing a generalized parity check matrix to be renormalized;

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Figure 8 is a graph comparing methods of evaluating error-correcting codes;

Figure 9 is a flow diagram of a renormalization group method according to the invention; and

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Figure 10 is a flow diagram of the method according to the invention for a graph with loops.

Detailed Description of the Preferred Embodiment

Renormalization Group Method

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Our invention evaluates error-correcting codes by using “real-space” renormalization group transformations. Our renormalization group (RG) method is adapted from techniques for the analysis of magnetic spin systems described by T. Niemeijer and J. M. J. van Leeuwen in “Phase Transitions and Critical Phenomena,” C. Domb and M. S. Green editors, Vol. 6, Academic Press, London, 1976. Renormalization groups have never been applied to error-correcting codes used to store and transmit data.

10

To evaluate the performance of a large but finite error-correcting code, we iteratively replace the large code with a slightly smaller code that has the same or nearly the same performance, i.e., decoding failure rate. In particular, at each step in our iterative method, we keep a Tanner graph and a set of probability variables p_{ia} and q_{ai} associated with messages transmitted by the nodes of the graph. For the purpose of describing the present invention, we call the combination of the Tanner graph and the p and q variables a “decorated Tanner graph.”

20

The basis of our RG method is the RG transformation by which we iteratively eliminate, i.e., “renormalize,” single nodes in the decorated Tanner graph, and adjust the remaining values of the p and q messages so that the smaller error-correcting code has a decoding failure rate as close as possible to the replaced code.

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Thus, with each renormalization step, the decorated Tanner graph representing our code shrinks by one node, until the graph is finally small enough that the performance of the error-correcting code can be determined in an efficient manner.

- 5 In contrast, prior art density evolution methods never change the number of nodes, or the way the nodes are connected to each other. In other words, their graphs are static, while our graphs change dynamically during renormalization. When our graph is small enough, e.g., when the number of remaining nodes is less than a predetermined threshold, the failure rate can readily be determined. In one
10 embodiment, we reduce the graph down to a single node.

Figure 9 shows the steps of the general method 900 according to our invention. For each selected “target” variable node i in our decorated Tanner graph for which the decoding failure rate b_i is to be determined, repeatedly renormalize a single node
15 from the graph, other than the target node, until a predetermined threshold 950 is reached. The threshold can either be expressed as a desired number of nodes, or a desired failure rate for the remaining node. Nodes are renormalized as follows.

- Measure 910 the “distances” between every other node and the “target” node i . The
20 distance between two nodes is the minimal number of nodes through which one passes to travel from one node to the other, i.e., the number of intervening nodes.

- If there are any “leaf” variable nodes, then renormalize 920 a leaf variable node
25 farthest from the “target” node. A node is a leaf node when it is connected to only one other node in the graph. Ties in distance can be broken randomly.

Otherwise, if there are no “leaf” variable nodes, then renormalize 930 a “leaf” check node farthest from the “target” node. Again, ties can be broken randomly.

Otherwise, if there are no “leaf” check nodes renormalize 940 a variable node from among those farthest from the target node that is directly connected to the fewest number of check nodes. Again, ties can be broken randomly.

The renormalization steps are iteratively repeated until the graph has been reduced down to the desired number of nodes. The details of these steps are described in greater detail below.

The above steps can be applied to as many target nodes as desired; for example, the average failure rate of every node in the graph can be determined. In a practical application, the renormalization is applied to target nodes representative of groups of like nodes. Then, it can be determine if the group of nodes is “strong” or “weak” with respect to its error-correcting capabilities. This information can then be used to improve the overall performance of the error-correcting code. For example, nodes in weak groups can be connected to more parity checks. Thus, the invention provides means for improving error-correcting codes in a structured manner.

The RG Transformation for a Decorated Tanner Graphs with no Loops

First, we consider loop-free Tanner graphs, and determine the RG transformations that are sufficient to give exact failure rates for such error-correcting codes. Then, we extend the RG transformations in order to obtain good approximate results for Tanner graphs with loops.

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We always initialize our decorated Tanner graph such that all $b_l = x$, $p_{ia} = x$, and all $q_{ai} = 0$. We are interested in the decoding failure rate b_i at the specified target variable node i . Our method 900 obtains b_i by repeatedly renormalizing nodes, one node at the time, as described above, other than the target variable node i itself.

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The first possibility is to renormalize the farthest "leaf" variable node that is connected to the target node i . Clearly, when that leaf node "vanishes," p_{ia} and q_{ai} are also discarded. We also renormalize all the q_{aj} variables leading out of the target node i to other variable nodes j . Our formulation for this renormalization is:

$$q_{aj} \leftarrow 1 - (1 - q_{aj})(1 - p_{ia}) \quad (37)$$

where the left arrow indicates that we replace the old value of q_{aj} with the new value. Notice that each renormalization of q_{aj} *increases* its value.

15

When we renormalize a "leaf" check node a that is only connected to a single variable node, we adjust the values of all the p_{ib} variables leading to other checks nodes b to which the target node i is attached. The renormalization group transformation is

$$p_{ib} \leftarrow p_{ib} q_{ai} \quad (38)$$

20

Notice that each renormalization of p_{ib} *decreases* its value. At the same time, we should also renormalize the b_i as follows:

$$b_i \leftarrow b_i q_{ai} \quad (39)$$

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25 The renormalization of step 940 is described in greater detail below for loopy graphs.

5 When only the "target" node i remains, we use the current value of node b_i as our RG prediction of the average failure rate. As stated above, these steps can be repeated for any number of target nodes.

Example

To better describe the RG method according to our invention, we give the following simple example. Recall the error-correcting code defined by the parity check matrix

$$A = \begin{pmatrix} 1 & 1 & 0 & 0 \\ 0 & 1 & 1 & 1 \end{pmatrix}. \quad (40)$$

15 We desire to determine the decoding failure rate at the second variable node b_2 . We initialize $p = p_{21} = p_{22} = p_{32} = p_{42} = x$, $q_{11} = q_{12} = q_{22} = q_{23} = q_{24} = 0$, and $b_2 = 0$.

As shown in Figure 3(a), we decorated the Tanner graph for this code. All of the variable nodes other than variable node 2 are leaf nodes, so we can renormalize any of them. According to our general method 900, we renormalize the node farthest from node 2, breaking ties randomly. If we select variable node 4, then we discard p_{42} and q_{24} and obtain new values $q_{22} = x$, and $q_{23} = x$ using equation (37).

The new reduced size decorated Tanner graph is shown in Figure 3(b). Next, we renormalize variable node 3. We discard nodes p_{32} and q_{23} , and renormalize node q_{22} to the value $1 - (1 - x)^2 = 2x - x^2$. The even smaller decorated Tanner graph is shown in Figure 3(c). Next we renormalize variable node 1. We discard nodes p_{11}

and q_{11} , and obtain yet smaller renormalized value $q_{12} = x$. The Tanner graph is shown in Figure 3(d). Next we renormalize check node 2. We can discard nodes p_{22} and q_{22} and obtain $p_{21} = b_2 = 2x^2 - x^3$ as shown for the Tanner graph in Figure 3(e). Finally we renormalize check node 1. This leaves us with only a single
 5 variable node, our original target node 2, and b_2 gets renormalized to its correct failure, $b_2 = 2x^3 - x^4$, as described above.

This example makes it clear why the RG method is exact for a codeword defined on a graph without loops. The RG transformations essentially reconstruct the
 10 density evolution equations of the prior art, and we know that density evolution is exact for such an error-correcting code. The advantage of our RG method is that it gives a much better approximation for error-correcting codes represented by bipartite graphs *with* loops. It is well understood, that good error-correcting codes, even of moderate size, will always have loops, mainly because loops provide
 15 redundancy without substantially increasing the size of the code.

The RG Method for a Graph with Loops

For an error-correcting code represented by a graph with loops, we eventually have
 20 to renormalize a node that is not a “leaf” node. Note that we never have to renormalize a non-leaf check node. To do this, we first collect all the check nodes a, b , etc. connected to the target node i . We discard $q_{ai}, q_{bi}, p_{ia}, p_{ib}$, etc. For any given check node attached to node i , e.g., check node a , we also collect all the other variable nodes j attached to node a , and renormalize the values of q_{aj} .

The renormalization of the q_{aj} variable can be done to varying degrees of accuracy. The simplest method uses equation (37) directly. The problem with this method is that the value of p_{ia} is always an over-estimate. Recall that p_{ia} *decreases* with every renormalization. Because we are renormalizing the i th node before it has become a
 5 leaf node, p_{ia} is not yet been fully renormalized, and is thus over-estimated.

Instead of using p_{ia} directly, we could use the value that it would have after we renormalized all the check nodes connected to it. That is, we could replace p_{ia} in equation (37) with an effective p_{ia}^{eff} given

$$p_{ia}^{\text{eff}} = p_{ia} \prod_{b \in N(i) \setminus a} q_{bi}. \quad (41)$$

On the other hand, we know that the values of the q_{bi} are under-estimates because they have not yet been fully renormalized either, so p_{ia}^{eff} also is an under-estimate. We could attempt to correct this mistake by going another level further. Before we
 15 estimate a p_{ia}^{eff} , we first re-estimate the q_{bi} which feed into it. Thus, we replace the p_{ia} in equation (37) with an effective p_{ia}^{eff} given by

$$p_{ia}^{\text{eff}} = p_{ia} \prod_{k \in N(i) \setminus a} q_{bi}^{\text{eff}}. \quad (42)$$

where q_{bi}^{eff} is in turn given by

$$q_{bi}^{\text{eff}} = 1 - (1 - q_{bi}) \prod_{k \in N(b) \setminus i} (1 - p_{kb}). \quad (43)$$

Putting all these together, we finally get the RG transformation

$$q_{aj} \leftarrow 1 - (1 - q_{aj}) \left(1 - p_{ia} \prod_{b \in N(i) \setminus a} \left[1 - (1 - q_{bi}) \prod_{k \in N(b) \setminus i} (1 - p_{kb}) \right] \right). \quad (44)$$

The RG transformation of equation (44) is worth describing in greater detail.

In Figure 4, we show the Tanner graph where variable node i is attached to three checks node a , b , and c , and check node a is attached to a variable node j . Check nodes b and c are connected to variable nodes labeled k , l , m , and n . We would like to know the probability q_{aj} that check node a sends variable node j an erasure message, taking into account the information that flows through variable node i .

We already have some previous accumulated probability q_{aj} that check node a sends to variable node j in an erasure message because of other nodes previously attached to check node a that have already been renormalized.

The new probability of an erasure message can be determined from a logical argument:

“ m_{aj} is an erasure it was already **or** if m_{ia} is an erasure

and

(m_{bi} **or** m_{kb} **or** m_{lb} are erasures) **and** (m_{ci} **or** m_{mc} **or** m_{nc} are erasures).”

Converting such a logical argument into an equation for probabilities is straightforward. If we have “ m_1 **and** m_2 ” for two messages in a logical argument, then we translates these terms to $(p_1 p_2)$ for the corresponding probabilities, while “ m_1 **or** m_2 ” translates to $(1 - (1 - p_1) (1 - p_2))$. Converting our full logical argument for Figure 4 into an equation for probabilities, enables us to recover an example of the RG transformation of equation (44).

We always have an RG transformation for q_{aj} correspond to the logic of the local neighborhood around the variable node i that we are renormalizing. In fact, the RG

transformation given in equation (44) is appropriate if the local neighborhood of node i is tree-like, but should be adjusted if there are loops in the local neighborhood.

- 5 For example, the graph in Figure 5 shows a case where a variable node k is attached to two check nodes b and c , which in turn are each attached to variable node i that is to be renormalized. Note that before check nodes b or c are renormalized, the probabilities p_{kb} and p_{kc} that variable node k sends out an erasure must be identical, because all renormalizations of p_{kb} and p_{kc} happen in tandem.

10 Our logic argument for whether check node a sends variable node j an erasure message is thus:

“ m_{aj} is an erasure if it was already **or** if (m_{ia} is an erasure)
and

(m_{kb} is an erasure) **or** (m_{bi} **and** m_{ci} are erasures))).”

15 At this stage in the renormalization process, if m_{kb} is an erasure, then m_{kc} must be as well. Converting our logic argument into an RG transformation, we get

$$q_{aj} \leftarrow 1 - (1 - q_{aj})(1 - p_{ia}(1 - (1 - p_{kb})(1 - q_{bi}q_{ci}))). \quad (45)$$

20 The step for renormalizing a non-leaf variable node i can be made increasingly accurate by increasing the size of the neighborhood around the node that was treated correctly. Naturally, as we increase the size of the neighborhood, we must pay for the increased accuracy with greater computation.

Figure 10 summarizes a method for renormalize a variable node that is not a leaf node, i.e., step 940 of Figure 9.

5 Step 1010 enumerates all check nodes a, b , etc. which are connected to the variable node to be renormalized and discards all the probability variables $p_{ia}, p_{ib}, q_{ai}, q_{bi}$, etc. between these check nodes and the variable node.

For each variable node j attached to one of the neighboring check nodes a , renormalize the value of q_{aj} according to the following sub-steps:

10

Step 1020 finds all check nodes and variable nodes in a local neighborhood to a predetermined distance from the variable node i to be renormalized. The distances are measured as described above.

15

Use a logical argument to determine 1030 which combinations of erasure messages cause the message from check node a to variable node j to also be an erasure.

Translate 1040 the logical argument into a RG transformation for q_{aj} as describe above.

20

Finishing the RG Transformation Exactly

As stated above, our RG method can always renormalize nodes until just the “target” node i is left, and then determine the decoding failure rate b_i . On the other
 5 hand, we can also renormalized a sufficient number of nodes to make an exact determination.

Sub For the purposes of describing the exact determination, we instead represent the error-correcting code by a Tanner graph of N nodes, and an associated an erasure probability x_i with each node i of the graph. This “erasure graph” is different than
 10 the decorated Tanner graphs 301-305 described above. The decoding failure rate can be determined exactly for an erasure graph, but the exact computation is only practical if the number of nodes in the erasure probability graph is small. We describe how to determine the decoding failure rate exactly for an erasure
 15 probability graph, and then describe how to convert a decorated graph according to the invention into an equivalent erasure graph.

To determine exactly the decoding failure rate of a given node i , we generate all 2^N possible received messages, ranging from the all-zeros message to the all-erasures
 20 message, and decode each of message using the BP decoder.

Each message has a probability

$$p = \prod x_i \prod (1 - x_j) \quad (46)$$

where the first product is over all nodes that are erased and the second product is
 25 over all nodes that are not erased. We determine b_i by taking the weighted average over all possible received messages of the probability that node i decodes to an

erasure. Because the complexity of the exact calculation is $O(2^N)$, we restricted ourselves to a small N , but nevertheless we can gain some accuracy by switching to an exact calculation after renormalizing sufficient nodes.

- 5 The one subtlety in the exact final calculation is that we now represent the error-correcting code by a Tanner graph and associated erasure probabilities x_i at each variable node i . In contrast, the general RG method uses just a decorated Tanner graphs. Fortunately, it is possible to convert a decorated Tanner graph into an erasure graph. Note that at each step of the RG method, all the probabilities q_{ai} leading out of the check node a are equal, i.e., $q_{ai} = q_a$, and all the probabilities p_{ia} leading out the variable node i are equal, i.e., $p_{ia} = p_i$.

- We can set all the q_a probabilities equal to zero by adding a new variable node k , with probability of $p_{ka} = q_a$, to node a in the erasure graph. When we are left with a decorated Tanner graph, such that all q probabilities are zero, and all p_{ia} probabilities coming out of each variable node are equal to p_i . We interpret p_i as the erasure probabilities of the variable nodes.

- Figure 6 shows an example of expanding a decorated Tanner graph 601 to an equivalent erasure Tanner graph 602 with erasure probabilities.

Extension to Generalized Parity Check Matrices

- Generalized parity check matrices define many of the modern error-correcting codes, such as turbo-codes, Kanter-Saad codes, and repeat-accumulate codes. In the generalized parity check matrix, additional columns are added to a parity check

matrix to represent “hidden” nodes. Hidden nodes have state variables that are not passed to other nodes, i.e., the states of the hidden nodes are “hidden.”. A good notation for the hidden state variables is a horizontal line above the corresponding columns. For example, we write

$$A = \begin{pmatrix} \overline{1} & 1 & 0 & 1 & 0 & 0 \\ 1 & 0 & 1 & 0 & 1 & 0 \\ 0 & 1 & 1 & 0 & 0 & 1 \end{pmatrix} \quad (47)$$

to indicate a code where the first variable node is a hidden node. To indicate that a variable node is a hidden node in our graphical model, we use an open circle rather than a filled-in circle. Such a graph, which generalizes Tanner graphs, is called a “Wiberg graph,” see N. Wiberg, “Codes and decoding on general graphs,” Ph. D. Thesis, University of Linköping,” 1996, and N. Wiberg et al., “Codes and iterative decoding on general graphs,” Euro. Trans. Telecomm, Vol. 6, pages 513-525, 1995.

Figure 7 shows a Wiberg graph representing an error-correcting code defined by the generalized parity check matrix or equation (47).

The generalization our RG method 900 to handle Wiberg graphs, we initialize the probabilities p_{ia} coming out of a hidden node to one, instead of at the erasure rate x , as we do for ordinary transmitted variable nodes. This reflects the fact that hidden nodes are always erased, while ordinary variable nodes are erased with a probability of x .

Comparison with Numerical simulations

We now describe some experimental predictions of our RG method. We define a (3,5) regular Gallager error-correcting code by a parity check matrix corresponding with $N = 60$ and $k=36$. That is, each of the 36 rows in the parity check matrix has five entries that are ones, and the rest are zeros, and each of the 60 columns has three entries that are zeros. There are no hidden nodes. No two parity checks share more than one variable node. This means that all local neighborhoods of nodes are tree-like. Therefore, we use the RG transformation (44) whenever we renormalize a non-leaf variable node. We renormalized nodes until we are left with seven nodes, and then finish the determination exactly.

We consider erasure rates x at intervals of 0.05 between $x = 0$ and $x = 1$. When we use the RG approximation, we average our decoding failure rates b_i over all nodes i to get an overall bit error rate. Our experiment includes a thousand trials at each erasure rate, while we decode according to the standard BP decoding method.

Figure 8 shows our experimental results. In Figure 8, the x-axis is the erasure rate, and the y-axis the bit error rate. Curve 811 is the prior art density evolution prediction, curve 812 the RG theoretical prediction, and the open circles 813 our experimental results. Figure 8 clearly shows that the density-evolution prediction has a threshold-like behavior and is completely incorrect for small or medium block-lengths, whereas the RG method according to the invention is not.

Extension to a Gaussian Noise Channel

We extend our RG method so that it can also be used with an additive white Gaussian noise (AWGN) channel. We do this by adapting a Gaussian

- 5 approximation to density evolution for the AWGN channel as described by Chung, et. al. in "Analysis of Sum-Product Decoding of Low-Density Parity-Check Codes Using a Gaussian Approximation," IEEE Trans. Info. Theory, Vol. 47, No.2, pages 657-670, 2001. We first describe that approximation.

- 10 In the AWGN channel, there are only two possible inputs, 0 and 1, but the output is a set of real numbers. If x is the input, then the output is $y = (-1)^x + z$, where z is a Gaussian random variable with zero mean and variance σ^2 . For each received bit i in the code, the log-likelihood ratio $m_i^0 = \ln(p(y_i|x_i = 0)/p(y_i|x_i = 1))$ determines the relative log-likelihood ratio that the transmitted bit i was a zero, given the received
- 15 real number is y_i .

The error-correcting code is defined by the generalized parity check matrices, as described above. The all-zeros codewords are transmitted, and the decoding process is the sum-product belief propagation process. In this decoding process,

- 20 real-valued messages are iteratively solved as functions of each other. The types of real-valued messages which are used are m_{ia} from variable nodes i to check nodes a ; and m_{ai} from check nodes a to variable nodes i .

See
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- 25 The messages m_{ia} are log-likelihood ratios by which the variable node i informs the check node a of its probability of being either a zero or a one. For example,

Sub 19 $m_{ia} \rightarrow \infty$ means that node i is certain it is a zero, while $m_{ia} = 1$ means that variable node i is signaling check node a that $\ln(p(x_i = 0)/p(x_i = 1)) = 1$. The messages m_{ai} are log-likelihood ratios interpreted as information from the check node a to the variable node i about the state of variable node i .

5

In sum-product decoding, the messages are iteratively solved according to the update rules:

$$m_{ia} = \sum_{b \in N(i) \setminus a} m_{bi} + m_i^0, \quad (48)$$

if node i is a hidden node, the m_i^0 is omitted, and

$$\tanh(m_{ai} / 2) = \prod_{j \in N(a) \setminus i} \tanh(m_{ja} / 2). \quad (49)$$

Sub 20 In the density evolution method for the AWGN channel, one considers the probability distributions $p(m_{ia})$ and $p(m_{ai})$ for the messages where the probability distribution is an average over all possible received blocks. A distribution $f(x)$ is called consistent if $f(x) = f(-x)e^x$ for all x . The consistency condition is preserved for the message probability distributions for all messages under sum-product decoding.

Sub 21 If the probability distributions $p(m_{ia})$ and $p(m_{ai})$ are approximated as Gaussian distributions, then the consistency condition means the means μ of these distributions are related to the variances σ^2 by $\sigma^2 = 2\mu$. This means that the message probability distributions can be characterized by a single parameter: their mean.

Thus, by making the approximation that the message probability distributions are Gaussians, the density evolution equations for the AWGN channel can be reduced to self-consistent equations for the means u_{ia} of the probability distributions of messages from variable nodes i to check nodes a , and the means v_{ai} of the

5 probability distributions of messages from check nodes a to variable nodes i . These equations are

$$v_{ia} = u^0 + \sum_{b \in N(i) \setminus a} u_{bi} \quad (50)$$

where u^0 is the mean value of m_i^0 , and is omitted for hidden nodes, and

$$\phi(u_{ai}) = 1 - \prod_{j \in N(a) \setminus i} (1 - \phi(v_{ja})), \quad (51)$$

10 where $\phi(x)$ is a function defined by

$$\phi(x) \equiv 1 - \frac{1}{\sqrt{4\pi x}} \int_{-\infty}^{\infty} \tanh \frac{u}{2} e^{-\frac{(u-x)^2}{4x}} du. \quad (52)$$

RG transformations for the AWGN channel

15 The density evolution equations (50) and (51) for the AWGN channel under the Gaussian approximation are analogs of the density evolution equations (4) and (3) for the BEC channel. Our RG procedure for the AWGN channel is substantially the same as for the BEC channel. The main difference is that we change the RG transformations.

20

Just as before, we construct a set of RG transformations which exactly reproduce the density evolution equations for a tree-like graph. We generate a decorated Tanner/Wiberg graph for the code by keeping u_{ai} and v_{ia} variables between each pair of connected nodes. The u_{ai} variables are initialized to infinity, while the v_{ia}

variables are initialized to u^0 , unless the i th node is a hidden node, in which case the v_{ia} are initialized to zero. We also introduce the variables h_i , analogous to b_i in the BEC, which are initialized like the v_{ia} variables.

- 5 If we renormalize a leaf check node a connected to a variable node i , then we find the other check nodes b attached to i and apply the RG transformations

$$v_{ib} \leftarrow v_{ib} + u_{ai} \quad (53)$$

and

$$h_i \leftarrow h_i + u_{ai}.$$

10 If we renormalize a leaf variable node i connected to a check node a , we find the other variable nodes j attached to check node a and apply the RG transformation

$$u_{aj} \leftarrow \phi^{-1}(1 - (1 - \phi(u_{aj}))(1 - \phi(v_{ia}))). \quad (55)$$

- 15 Note that with each renormalization of v_{ib} , the magnitude of v_{ib} increases, while with each renormalization of u_{aj} , and the magnitude of u_{aj} decreases.

When we renormalize a non-leaf variable node i which is connected to check nodes a, b , etc., we renormalize the variables like u_{aj} , where j is another variable node

- 20 connected to check node a . Just as for the BEC, we consider a local neighborhood of nodes around the node i . For example, if the neighborhood of check nodes connected to i and other variable nodes connected to those check nodes is tree-like, we use the RG transformation

$$u_{aj} \leftarrow \phi^{-1}(1 - (1 - \phi(u_{aj}))(1 - \phi(v_{ia}^{\text{eff}}))), \quad (56)$$

- 25 where

$$v_{ia}^{\text{eff}} = v_{ia} + \sum_{b \in N(i) \setminus a} \phi^{-1} \left(1 - (1 - \phi(u_{bi})) \prod_{k \in N(b) \setminus i} (1 - \phi(v_{kb})) \right). \quad (57)$$

The RG method proceeds as in the BEC case, until the determination of the bit error rate. For the AWGN channel, it is normally inconvenient to stop the RG method before renormalizing all the way down to the “target” node, because it is not simple to do an exact computation even with just a few nodes in the graph.

When we have renormalized all but our target node i , we are left with a final renormalized value of h_i . The Gaussian approximation tells us that the probability distribution for the node i being decoded as a zero is a Gaussian with mean h_i and variance $2h_i$. Decoding failures correspond to those parts of the probability distribution which are below zero. Thus, our prediction for the bit error rate (ber_i) at node i is

$$\text{ber}_i = \frac{1}{\sqrt{8\pi h_i}} \int_{-\infty}^0 e^{-\frac{(x-h_i)^2}{4h_i}} dx. \quad (58)$$

Generating Error-Correcting Codes

Given that the density evolution method has been used as a guide to generate the best-known practical error-correcting codes, we can generate even better codes using our RG method. With the RG method according to our invention, we can input a code defined by an arbitrary generalized parity check matrix, and obtain as output a prediction of the bit error rate at each node.

We can use this output in an objective function for a guided search through the space of possible improved code. For example, we can try to find a $N = 100$ blocklength, transmission rate $1/2$ code with no hidden states that achieves a bit error rate of less than 10^{-4} at the smallest possible signal-to-noise ratio for the

5 AWGN channel. We do this by iteratively evaluating codes of the correct blocklength and rate, using our RG method, and using any known search techniques, e.g., greedy descent, simulated annealing, genetic process, etc. to search through the space of valid parity check matrices.

10 Because we directly focus on the correct measure of merit, i.e., the bit error rate itself, rather than the threshold in the infinite block-length limit, the search according to the invention improves on the results obtained using the prior art density evolution process. We can guide the search because we have information about the bit error rate at every node. For example, it might make sense to

15 “strengthen” a weak variable node with a high bit error rate by adding additional parity check nodes, or we can “weaken” strong nodes with a low bit error rate by turning the weak nodes into hidden nodes, thus increasing the transmission rate.

20 On the other hand, determining the bit error rate of every node slows down a search. It may be worthwhile, at least at least for large block-lengths, to restrict oneself to those codes for which there are only a small number of different classes of nodes, defined in terms of the local neighborhoods of the nodes. Most of the best-known codes are of this type. Rather than determining the bit error rate for every variable node, we can determine the bit error rate for just one representative

25 node of each class of variable nodes.

For example, for a regular Gallager code, each node has the same local neighborhood, so any node can be selected as a representative of all the nodes in the neighborhood. The error made in this method is estimated by comparing bit error rates of different nodes of the same class. For actual finite-sized regular

- 5 Gallager codes, we find that the RG method gives very similar predictions for each of the nodes, so that the error made by just considering a single variable node as a representative of all of them is quite small.

- 10 Although the invention has been described by way of examples of preferred embodiments, it is to be understood that various other adaptations and modifications may be made within the spirit and scope of the invention. Therefore, it is the object of the appended claims to cover all such variations and modifications as come within the true spirit and scope of the invention.